

# Low Driving Voltage Lithium Niobate Metasurface Electro-Optical Modulator Operating in Free Space

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**Abstract:** A simple configuration of only  $\lambda/9$  thick 2D metallic grating embedded within an electro-optic (EO) material (lithium niobate for instance) is proposed and theoretically studied to act as an EO modulator. On the first hand, this grating is used as an interdigitated comb to apply a very high and spatially periodic modification of the electrostatic field. On the other hand, the grating is designed to exhibit a Fano-like resonance in the NIR spectral range. This resonance is used to confine the electromagnetic field inside the EO material leading to an intrinsic enhancement of the EO effect. Extensive numerical simulations are performed to optimize the geometry in agreement with technological fabrication constraints. We achieved a local field factor of 24.5 leading to a local index modification  $\Delta n$  as large as 1 for 1 V applied voltage. This allows a modulation sensitivity of 14.35 nm/V (2000 times larger than the state of the art) together with a resonance depth of 60% and a driving voltage of only 75 mV opening the way to the fabrication of ultra-thin low driving voltage EO devices.

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## 1. Introduction

Electro-optical (EO) modulators required to convert electrical signals into optical informations are crucial and in very high demand for most optical-based devices and telecommunication systems [1, 2]. In this context, lithium niobate ( $\text{LiNbO}_3$ ) is known as one of the best candidate for EO devices due to its important physical properties, notably its high EO coefficients [3–5]. It offers a wide transparency band, which opens the range to applications from visible to mid-infrared. The dominance of  $\text{LiNbO}_3$  in EO technologies is related to its higher Curie temperature compared to other EO materials, which makes it robust to temperature variation and harsh environments.

The efficiency of an EO modulator is linked to the electric voltage needed to drive the modulator and to the footprint of the device that can be assessed with the  $V_d \cdot L$  figure of merit, where  $V_d$  is the driving voltage and  $L$  the active length. In general, commercial EO modulators made of  $\text{LiNbO}_3$  and based on a Mach-Zehnder interferometer (MZI) present a half wave voltage between 3 V [6] and 1 V [7] with an active length larger than one centimeter.

After the emergence of thin-film lithium niobate (TFLN) [8, 9], a new era in photonics technology has begun, leading to EO devices with reduced footprint and driving voltage. In the last couple of years, the progress remains improving in order to conceive EO modulators with lower half wave voltage ( $V_\pi$ ) and larger bandwidth taking advantage of TFLN [10–12]. In [10], they achieved a 1 V driving voltage and 110 GHz bandwidth in a TFLN-modulator that works in dual-polarization, enabling a record single-wavelength 1.96 Tb/s net data rate. However, Weiss *et al.* [12] proposed a plasmonic nanoscale metasurface array coupled with a TFLN and reach a wavelength tunability of 3 nm for a 50 V peak-to-peak voltage difference. Generally, the voltage-length product ( $V_\pi \cdot L$ ) of the TFLN modulator is 1.5 to 3 V·cm [13, 14], which is considerably lower than that of commercial LN modulators (15 V·cm) [1] at telecom wavelengths.

45 Photonic Crystals (PhC) can exalt EO interactions and gain even more than an order of  
46 magnitude on the figure of merit compared to MZ-based architectures [15]. Moreover, PhC  
47 architecture can thereby shrink down the device footprint which reduces its capacitance and the  
48 required switching voltage. The emergence of PhC-based devices has increased the integration  
49 of photonic devices to meet the requirements of embedded technologies and become competitive  
50 with electronic devices.

51 A LiNbO<sub>3</sub> PhC modulator was proposed since 2007 to enhance the EO effect by locally  
52 confining the electromagnetic field when a Bloch mode is excited within the PhC [16]. This  
53 becomes relevant when the group velocity of this Bloch mode is very small. In this case, the time  
54 interaction between light and matter grows significantly leading to exacerbate all the intrinsic  
55 non-linear coefficients of the considered material, among others, the second order susceptibility  
56 term  $\chi^{(2)}$  involved in the EO effect. Such enhancement was experimentally [17] verified  
57 and theoretically explained. Thereby, a new generation of EO modulators [18], pyroelectric  
58 detectors [19] or Second Harmonic Generation (SHG) devices [20] were proposed, fabricated and  
59 characterized validating the principle of this enhancement. Meanwhile, dark Fano resonances,  
60 also known by optical bound state in the continuum (BIC), were also explored in order to exacerbate  
61 the EO effect [21]. It provides an excellent electromagnetic field confinement essential to enhance  
62 the EO effect surpassing a conventional guided resonance [22].

63 Here, we proposed a simple Ag-LiNbO<sub>3</sub> configuration based on the excitation of a Fano-like  
64 resonance at  $\lambda_{\text{res}} = 1530$  nm insuring the electromagnetic field confinement required to enhance  
65 the host medium EO coefficient. The geometry of the structure is inspired by the one studied  
66 in [23] where a miniaturized acousto-optic modulator based on sub-wavelength structures  
67 presenting simultaneously a photonic and phononic resonance has been proposed and studied  
68 demonstrating a strong optical modulation at near infrared wavelengths.

## 69 2. Proposed Structure

70 Unlike most common modulators, we propose a configuration operating in the  $\Gamma$  direction  
71 (out-of-plane illumination at normal incidence), which allows us to illuminate the structure in  
72 free space and overcome the optical losses that occur by an in-plane illumination such as the  
73 injection losses and the propagation losses, as it is the case for MZI-based modulators. Our EO  
74 modulator, as shown in Fig.1(a), is based on a simple configuration of a 2D metallic grating of  
75 thickness  $h = \lambda/9$ , embedded in the LiNbO<sub>3</sub> in order to take advantage of the confinement of the  
76 electromagnetic field induced by our Fano resonance. On one hand, this grating is used as an  
77 interdigital comb to apply a very high and spatially periodic variation of the electrostatic field  
78 leading to an enhancement of the EO effect and improvement of the EO sensitivity. On the other  
79 hand, the grating is designed to optically exhibit a Fano type resonance in the NIR spectral range.  
80 Our proposed structure relies on a 1D PhC structure with a grating lattice  $p$ . The unbalance  
81 between the two LiNbO<sub>3</sub> cavities with two different widths  $W_1$  and  $W_2$  creates a mismatch  
82 between the volume of the cavities and allow us to generate the well-known  $\pi$ -resonance also  
83 known as phase-resonance. The origin of the optical resonance appearing on the transmission  
84 spectrum through the structure has been discussed in [24]. In order to take advantage of the larger  
85 EO coefficient of the LiNbO<sub>3</sub>, the applied voltage and electromagnetic field must be oriented  
86 parallel to the crystallographic  $Z$ -axis of the LiNbO<sub>3</sub>, which requires an  $X$ - or  $Y$ -cut LiNbO<sub>3</sub>  
87 slab. In the following, we will consider an  $X$ -cut LiNbO<sub>3</sub> structure and assume the LiNbO<sub>3</sub>  
88 extraordinary refractive index  $n_e = 2.2472$  to be constant in the spectral range of interest.

89 In order to adapt the geometry, extensive FDTD (Finite Difference Time Domain) simulations  
90 with a custom code were performed to obtain resonance in the NIR spectral range with the  
91 highest Q-factor and the largest resonance depth (RD) corresponding to a maximum modulation  
92 amplitude. A uniform 2D mesh with a mesh size of  $\Delta x = \Delta y = 2$  nm is applied to describe  
93 the structure. We consider an infinitely periodic structure in the  $x$  direction by applying the

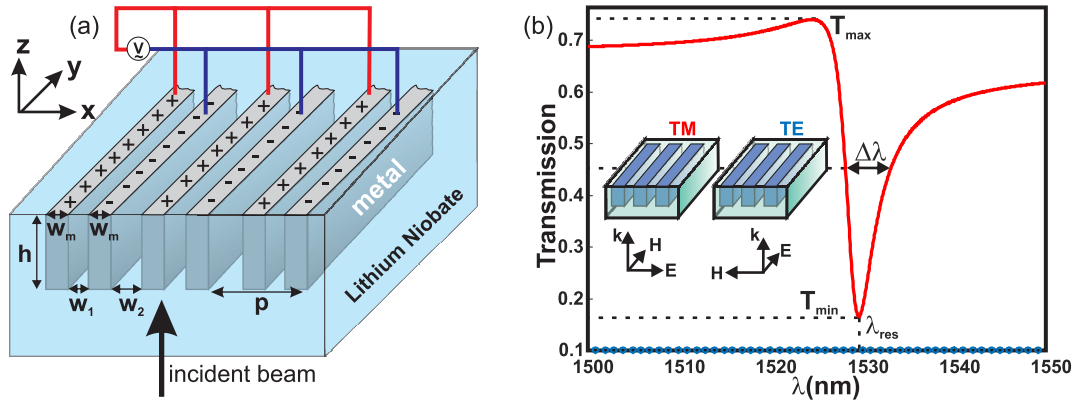


Fig. 1. (a) Schematic of the proposed structure showing its geometrical parameters and (b) transmission spectra for TE polarization presented by blue circles and by solid red line for the TM-polarized incident plane waves. The geometrical parameters are:  $h = 170\text{nm}$ ,  $p = 650\text{nm}$ ,  $W_1 = 150\text{nm}$ ,  $W_m = 120\text{nm}$ . The refractive index of  $\text{LiNbO}_3$  is fixed to  $n_e = 2.2472$  and the metal is silver.

94 Bloch periodic boundary conditions, and perfectly matched layer (PML) conditions are used  
 95 as absorbing boundary conditions to numerically truncate the substrate and superstrate media.  
 96 Approximately 1 million iterations are considered for a single calculation to achieve the stopping  
 97 criteria (steady states). All geometrical parameters were varied taking into account the fabrication  
 98 constraints (aspect ratio  $AR = h/\min(W_m, W_1, W_2)$  smaller than 2 and slit width larger than  
 99  $80\text{nm}$ ). The geometrical parameters of the optimized structure are:  $h = 170\text{nm}$ ,  $p = 650\text{nm}$ ,  
 100  $W_1 = 150\text{nm}$ ,  $W_m = 120\text{nm}$ . Different metals were also studied (Ag, Al, Pt, Au) for the electrodes,  
 101 showing better results with silver. The latter presents the lowest optical absorption losses. Its  
 102 dielectric properties are adapted to [25] through a Drude critical points model [26].

103 Fig.1(b) shows a typical transmission spectrum of light through the structure for the two  
 104 linear polarization states (TE and TM). As expected, the slit structure behaves as a metallic  
 105 grid polarizer with axis perpendicular to the slit direction. Consequently, only the  $Ox$  incident  
 106 polarization can be transmitted. The resonance depth (RD) and the quality factor (Q-factor) are  
 107 defined from this spectrum by:  $RD = \frac{T_{\max} - T_{\min}}{T_{\max}}$  and  $Q = \frac{\lambda_{\text{res}}}{\Delta\lambda}$ . Their values, respectively 77%  
 108 and 318 for the resonance at  $\lambda_{\text{res}} = 1529.6\text{ nm}$ , are above the experimentally required threshold  
 109 values.

### 110 3. Results and Discussions

111 In order to enhance the EO effect in  $\text{LiNbO}_3$ , the electric field must overlap with the optical mode  
 112 in the  $\text{LiNbO}_3$ , in other words, the electric field and the electromagnetic field has to be confined  
 113 in the  $\text{LiNbO}_3$  substrate leading to a dielectric mode and to the enhancement of the EO coefficient  
 114  $r_{33}$  by a factor  $F_{\text{opt}}$  as described in equation 1, the expression of the local EO effect [16]:

$$\Delta n_e(x, z) = -\frac{1}{2} n_e^3 r_{33} F_{\text{opt}}^2(x, z) E_{\text{es}}(x, z), \quad (1)$$

115 where  $F_{\text{opt}}(x, z)$  is the local optical field factor defined by

$$F_{\text{opt}}(x, z) = \frac{|\vec{E}(x, z)|_{\text{structure}}}{|\vec{E}|_{\text{bulk}}}, \quad (2)$$

116 where  $|\vec{E}|_{\text{bulk}}$  is the amplitude of the homogeneous E-field in the bulk LiNbO<sub>3</sub> substrate (without  
 117 PhC) and  $|\vec{E}(x, z)|_{\text{structure}}$  is the local amplitude of the E-field in the studied PhC structure.  
 118 To confirm the existence of a dielectric mode and then derive the optical field factor  $F_{\text{opt}}$ , we  
 119 calculate the electric field distribution at the resonance wavelength using a 2D-FDTD custom  
 120 code. As shown in Fig.2(a), the normalized electric field intensity reaches its maximum at the  
 121 corners of the electrodes and inside the LiNbO<sub>3</sub> between the two electrodes leading to a dielectric  
 122 mode, which is consistent with the conditions required to enhance the EO effect. Based on the  
 123 optical field factor definition given in Eq.2, we reach a maximum of  $F_{\text{opt}}^2(x, z) = 1100$ , which  
 124 corresponds to a variation of the refractive index in the PhC structure up to 1100 times higher  
 125 than that of the bulk structure. The normalization has been established with the same FDTD  
 126 code and the same configuration of the calculation window without the nanostructure (periodic  
 127 metallic grid). However, as described in Eq.1, the EO sensitivity depends on  $F_{\text{opt}}(x, z)$  and  
 128 electrostatic field distribution through  $E_{\text{es}}(x, z)$ .

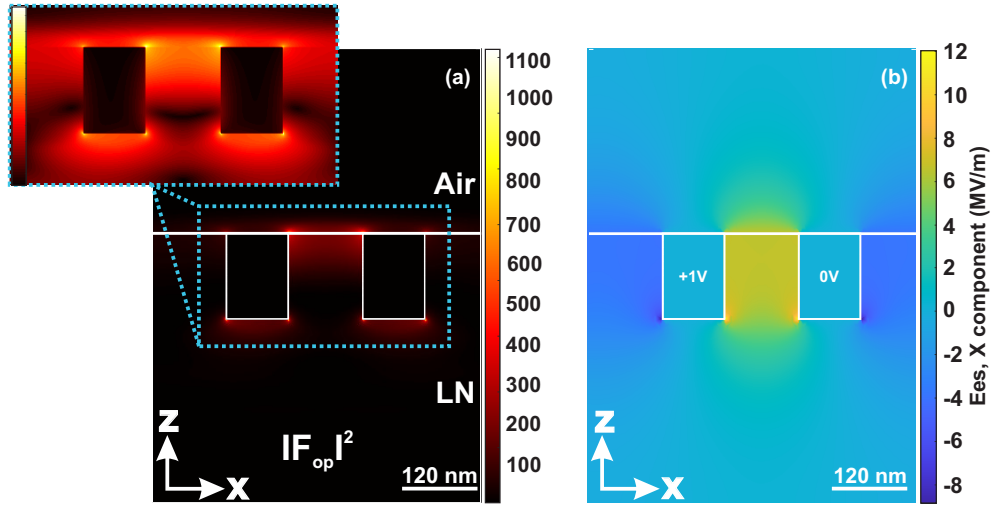


Fig. 2. (a) Spatial distribution of the normalized electric field intensity (or  $F_{\text{opt}}^2(x, z)$ ) at  $\lambda_{\text{res}} = 1529.6$  nm. The white lines correspond to the structure edges. The zoom-in made over the highlighted rectangular zone on the left shows the electric field intensity to the power 0.4 to enhance the color contrast and presents clearly the electric field distribution in this zone. (b) Electrostatic field distribution (only the x-component) when a potential difference of 1V is applied between two consecutive electrodes.

129 After the estimation of  $F_{\text{opt}}$  distribution, we moved to the calculation of the electrostatic field  
 130 distribution  $E_{\text{es}}(x, z)$  in order to complete the missing values for the estimation of the EO effect  
 131 of the proposed structure.

132 As previously explained, the applied electrostatic field  $E_{\text{es}}(x, z)$  is polarized along the x-  
 133 directions which corresponds to the crystallographic Z-axis for a X-cut LiNbO<sub>3</sub> wafer in order to  
 134 benefit from the highest EO coefficient of LiNbO<sub>3</sub> ( $r_{33}$ ). For this reason, we will only consider  
 135 in our calculations the X-components of  $E_{\text{es}}$  since it will involve the highest EO coefficient of  
 136 LiNbO<sub>3</sub> ( $r_{33}$ ). Similarly, we will only focus on the electrostatic field confined inside the LiNbO<sub>3</sub>  
 137 and not in the air, since the EO effect occurs only in the EO material (LiNbO<sub>3</sub> in our case).  
 138 Indeed, no electrostatic field apart from the one in LiNbO<sub>3</sub> will contribute to the EO effect.

139 The same structure was modeled to solve the Poisson equation when a voltage of 1 V is applied  
 140 between the two consecutive electrodes. Thus, COMSOL multiphysics software was used to  
 141 this end by integrating periodic boundary conditions in the x-direction and Dirichlet boundary

142 conditions for the up and down limits of the calculation window. In addition, a very fine mesh,  
 143 based on Delaunay triangulation, was applied in order to minimize the numerical errors when  
 144 interpolating the  $E_{es}(x,z)$  values over the regular square grid used in the FDTD simulations. The  
 145 x-component of the electrostatic field presented in Fig.2(b) is showing a maximum value between  
 146 the two electrodes and in the electrode corners immersed in  $\text{LiNbO}_3$ . It varies between  $12 \times 10^6$   
 147 V/m and  $-8.2 \times 10^6$  V/m, values that are slightly larger than

$$\frac{V}{W_1} = \frac{1}{150 \times 10^{-9}} = 6.67 \times 10^6 \text{ V/m}, \quad (3)$$

148 and

$$\frac{-V}{W_2} = \frac{-1}{260 \times 10^{-9}} = -3.85 \times 10^6 \text{ V/m}, \quad (4)$$

149 respectively. This difference comes from the presence of electrode corners ( $90^\circ$ ) that induces an  
 150 enhancement of the electric field due to a well-known "antenna effect" that was widely studied  
 151 both theoretically [27, 28] and experimentally [29, 30].

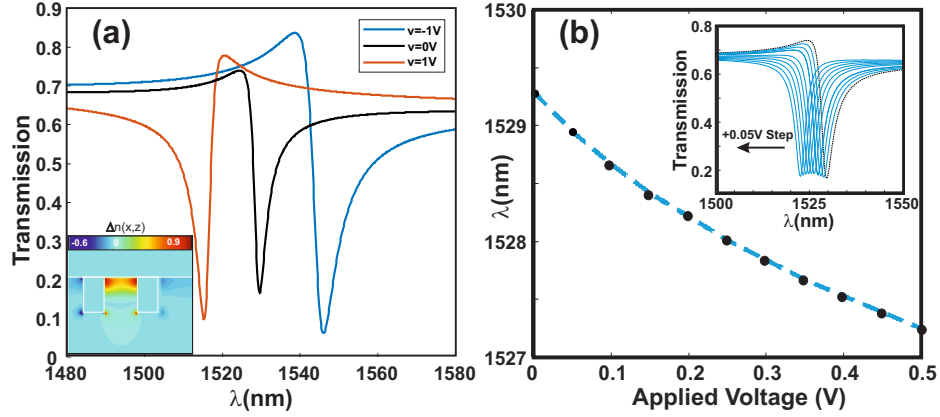


Fig. 3. (a) Shift of the resonance induced by an applied voltage of 1 V (in red) and  $-1$  V (in blue). In black, the transmission spectrum of structure before applying an electric voltage. Inset: the local refractive index distribution for 1 V applied voltage. (b) The evolution of the resonance wavelength with respect to the applied voltage. Inset in blue the transmission spectra of the structure for different electric voltage going from 0 V to 0.5 V with a step of 0.05 V and in black the transmission spectrum of the structure for 0 V.

152 To highlight the mechanism of modulation, we started by estimating the sensitivity of the  
 153 structure as response of an applied voltage. To do so, we compute the electrostatic field distribution  
 154 for an applied voltage of 1 V and  $-1$  V using COMSOL and inject these values with the optical  
 155 field factor distribution into Eq.1 in order to calculate the resonance shift by FDTD. In Fig.3(a),  
 156 we present the calculated three transmission spectra pointing out the resonance shift induced by 1  
 157 V (in red) and  $-1$  V (in blue) applied voltage. More details about the resonances are summarized  
 158 in Table 1.

159 For just 1 V applied voltage, the resonance (in red) shift to 1515.293 nm with a Q-factor of  
 160 252 and a RD of 87.3%. However, with a  $-1$  V applied voltage, the resonance (in blue) shift  
 161 to 1546.02 nm with a Q-factor of 135 and a RD of 92.4%. Despite the larger shift with a  $-1$  V  
 162 applied voltage, the decrease in the Q-factor of the resonance is directly linked to a lower optical  
 163 field factor compared to the other resonance (with an applied voltage of 1 V). In the following,

Table 1. Details about resonances in Fig.3(a).

$\Delta V$	$\lambda_{\text{res}}$	Q-Factor	RD
0 V	1529.652 nm	318	77.3%
1 V	1515.293 nm	252	87.3%
-1 V	1546.02 nm	135	92.4%

164 the modulation performance of the structure will be studied only with a positive applied voltage  
 165 so we can benefit as much as possible from the EO enhancement of the proposed structure.

166 A sensitivity of 14.35 nm/V is estimated, which is 2000 times higher than the one recently  
 167 achieved by Wang *et al.* [31]. The shift of the resonance wavelength with respect to the  
 168 applied voltage is presented in Fig.3(b). Similarly, every point is computed by a couple of  
 169 COMSOL-FDTD simulations. The trend of the curve with respect to the applied electric voltage  
 170 is quasi-linear.

171 To end up with the modulation performance of the structure, we present in Fig.4 the optical  
 172 response of the structure with respect to the applied voltage. We considered a light detector  
 173 operating at the resonance wavelength, and we deduce the transmitted light for each applied  
 174 voltage. We clearly see that we reach a permanent state for an applied voltage  $V=0.2$  V. This  
 175 means that our structure requires a driving voltage ( $V_d$ ) as low as 0.2 V in order to go from the  
 176 ON to the OFF states. In order to improve the modulation of the structure, one can consider the  
 177 highlighted red zone where the transmission variation as function of the applied voltage is linear.  
 178 We conclude with a minimized driving voltage  $V_d = 0.075$  V and an RD = 60% instead of  $V_d =$   
 179  $0.2$  V and an RD = 80% if we consider the complete modulation zone presented in Fig.4.

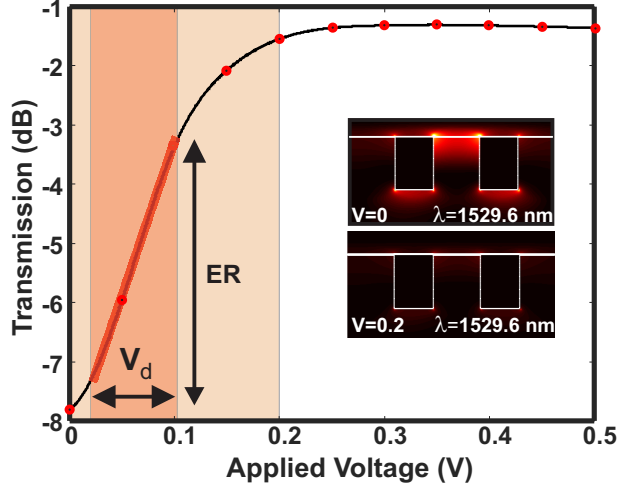


Fig. 4. The optical response of the structure with respect to the applied voltage. Inset:  
 the electric field amplitude at the ON and OFF state for  $V=0$  V and 0.2 V, respectively.

180 In order to be more realistic, we study the effect of a finite structure with a finite number of  
 181 period on its optical response. For this purpose, 2D-FDTD simulations are performed to evaluate  
 182 the normalized zero-order transmission spectra of the structure with a finite number of periods  
 183 and the previously determined parameters. The Bloch periodic boundary condition are replaced  
 184 by perfectly matched layer boundary conditions in order to simulate the finite structure with a  
 185 uniform mesh size of  $\Delta_x = \Delta_z = 10$  nm in the  $x$  and  $z$  directions instead of 2 nm in order to speed

186 up our simulations. The larger mesh size considered in our simulations will induce a shift of the  
 187 resonance wavelength according to [32]. The structure is illuminated by a Gaussian beam whose  
 size is fixed at  $100 \mu m$ .

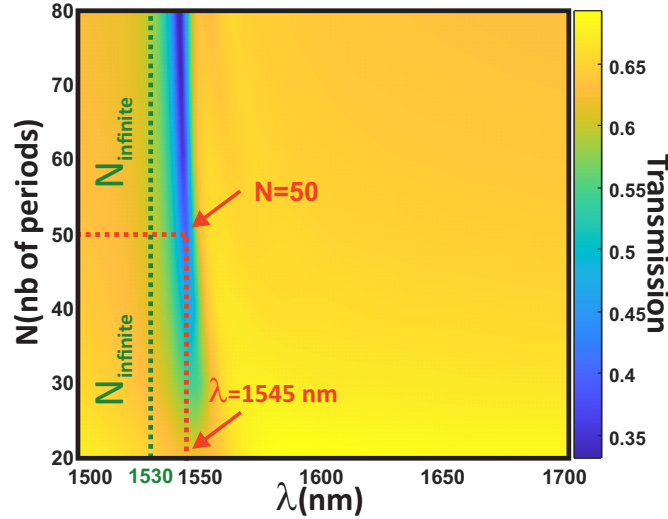


Fig. 5. Normalized transmission with respect to the number of periods. The dashed green line indicates the resonance position for an infinitely periodic structure, while the dashed red line refers to the resonance position, for 50 periods, where the RD and the Q factor of the resonance become sufficient for modulation.

188 The 2D-FDTD simulations are conducted by varying the number of periods, going from 20  
 189 periods to 80 periods. The result is presented in details in Fig.5 and table 2. On one hand,  
 190 Fig.5 presents the evolution of the transmission with respect to the number of period. Going  
 191 from 20 periods to 80 periods, the resonance wavelength shift from 1549 nm to 1543 nm. This  
 192 is reasonable since it converges to the resonance position of the infinitely periodic structure  
 193 denoted by the green dashed line. Admitting that a minimum RD of 40% is necessary to ensure  
 194 a proper modulation, we note that starting from 50 periods, the resonance properties become  
 195 sufficiently relevant to ensure the good modulation thanks to its significant RD and Q factor. This  
 196 corresponds to a footprint of the structure as small as about  $L=32.5 \mu m$ . On the other hand, the  
 197 table 2 shows the characteristics of the resonances for each number of periods. Similarly to what  
 198 is presented in Fig.5, starting from 50 periods, the variation of the resonance properties becomes  
 199 very small, revealing the convergence to almost stable values.  
 200

Table 2. Details of the resonance properties with respect to the number of periods N.

N	$\lambda_{res}$	Q-Factor	RD
20	1548.80 nm	56.6	10.26%
30	1547.08 nm	126.3	19.6%
40	1546.04 nm	164.2	30.1%
50	1545 nm	217.47	39.28%
60	1544.76 nm	268.84	44.98%
70	1544.16 nm	273.73	47.59%
80	1543.65 nm	273.97	49.73%

201 In order to compare the performance of our modulator with other modulators available in the  
 202 literature, we rely on the voltage-length product of the modulator expressed by the product  $V_d \cdot L$ ,  
 203 where  $V_d$  is the driving voltage of our structure and  $L$  the active length of the structure. If we  
 204 consider a finite structure with 60 periods ( $L = 39 \mu\text{m}$ ) and  $V_d$  0.075 V or 0.2 V, we reach a  
 205 voltage-length product of about  $2.925 \times 10^{-4} \text{ V}\cdot\text{cm}$  or  $7.8 \times 10^{-3} \text{ V}\cdot\text{cm}$ , respectively. In both  
 206 cases, our voltage-length product is more than two order of magnitude smaller than other LN EO  
 modulators [13, 15, 33]. In addition, a comparison table was provided in table 3 to benchmark all

Table 3. Comparison of modulator metrics: switching voltage, voltage-length product (V-L), ER and modulation tuning.

Type	Switching Voltage (V)	V-L (V·cm)	ER (dB)	Modulation Tuning	Reference
MZI	1.4	2.8	30	N/A	[13]
MZI	2.3	2.3	30	N/A	[13]
MZI	4.4	2.2	30	N/A	[13]
Micro-Ring	9	1.8	10	7 pm/V	[31]
Metasurface	100	N/A	4.7	20 nm/V (GMR)	[34]
Our structure	0.075/ 0.2	$2.925 \times 10^{-4} / 7.8 \times 10^{-3}$	4.7/ 6.7	14.35 nm/V	This work

207 relevant metrics.  
 208

#### 209 4. Conclusion

210 Our theoretical studies reveal the opportunity to develop an EO modulator with a minimized  
 211 driving voltage  $V_d = 75 \text{ mV}$  and a reduced footprint of  $L = 39 \mu\text{m}$  based on the well-known  
 212 EO effect in  $\text{LiNbO}_3$ , Pockels effect. We achieved a high sensitivity up to 14.35 nm/V and a  
 213 minimized voltage-length product of about  $2.925 \times 10^{-4} \text{ V}\cdot\text{cm}$  with our 2D metasurface-based  
 214 proposed structure. We reach a new horizon since, for the best of our knowledge, the deduced  
 215  $V_d \cdot L$  is the lowest in the literature and in the market.

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#### 221 Disclosures

222 The authors declare no conflicts of interest.

#### 223 Data availability

224 Data underlying the results presented in this paper are not publicly available at this time but may  
 225 be obtained from the authors upon reasonable request.

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